

TSDT14 Signal Theory

Lecture 6

Saturation, Quantization and Poisson Processes

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Half-Wave Rectifier

From the table:

$Y(t)$	$r_Y(\tau)$
$\begin{cases} X, & X \geq 0, \\ 0, & X < 0. \end{cases}$	$\begin{aligned} & \frac{r_X(\tau)}{4} + \frac{1}{2\pi} \left[\sqrt{r_X^2(0) - r_X^2(\tau)} + r_X(\tau) \arcsin\left(\frac{r_X(\tau)}{r_X(0)}\right) \right] \\ & = \frac{r_X(0)}{2\pi} + \frac{r_X(\tau)}{4} + \frac{r_X^2(\tau)}{4\pi r_X(0)} + \dots \end{aligned}$

Complete Maclaurin expansion:

$$r_Y(\tau) = \frac{r_X(0)}{2\pi} + \frac{r_X(\tau)}{4} + \frac{r_X^2(\tau)}{4\pi r_X(0)} + \sum_{n=2}^{\infty} \frac{(2n-3)!! \cdot r_X(0)}{2\pi \cdot (2n-1) \cdot (2n)!!} \left(\frac{r_X(\tau)}{r_X(0)}\right)^{2n}$$

$n!!$ is semi-factorial (product of every second positive integer):

$$6!! = 2 \cdot 4 \cdot 6 = 48 \quad \text{and} \quad 7!! = 1 \cdot 3 \cdot 5 \cdot 7 = 105.$$

More Non-Linearities

From
 Tables & Formulas,
 Page 15.

Situation

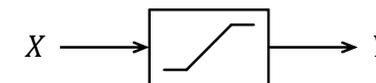
The input to a momentary non-linearity is a Gaussian process $X(t)$ with mean $m_X = 0$ and auto-correlation function $r_X(\tau)$.

Identities

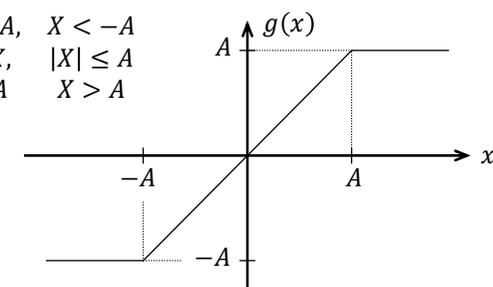
$Y(t)$	$r_Y(\tau)$
$X^2(t)$	$2r_X^2(\tau) + r_X^2(0)$
$X^3(t)$	$6r_X^3(\tau) + 9r_X^2(0)r_X(\tau)$
$X^4(t)$	$24r_X^4(\tau) + 72r_X^3(0)r_X^2(\tau) + 9r_X^4(0)$
$X^5(t)$	$120r_X^5(\tau) + 600r_X^4(0)r_X^3(\tau) + 225r_X^4(0)r_X(\tau)$
$\begin{cases} X, & X \geq 0, \\ 0, & X < 0. \end{cases}$	$\begin{aligned} & \frac{r_X(\tau)}{4} + \frac{1}{2\pi} \left[\sqrt{r_X^2(0) - r_X^2(\tau)} + r_X(\tau) \arcsin\left(\frac{r_X(\tau)}{r_X(0)}\right) \right] \\ & = \frac{r_X(0)}{2\pi} + \frac{r_X(\tau)}{4} + \frac{r_X^2(\tau)}{4\pi r_X(0)} + \dots \end{aligned}$
$\text{sgn}(X)$	$\frac{2}{\pi} \arcsin\left(\frac{r_X(\tau)}{r_X(0)}\right)$

The corresponding relations hold for time-discrete processes as well.

Saturation

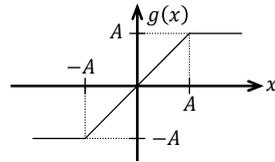


$$Y = g(X) = \begin{cases} -A, & X < -A \\ X, & |X| \leq A \\ A, & X > A \end{cases}$$



Saturation PDF

$$Y = g(X) = \begin{cases} -A, & X < -A \\ X, & |X| \leq A \\ A, & X > A \end{cases}$$

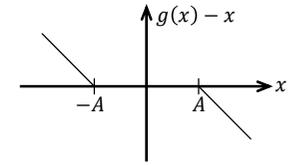


$$p_1 = \Pr\{X < -A\} = \int_{-\infty}^{-A} f_X(x) dx \quad p_2 = \Pr\{X > A\} = \int_A^{\infty} f_X(x) dx$$

$$\tilde{f}_Y(y) = f_X(y) \cdot \text{rect}\left(\frac{y}{2A}\right) = \begin{cases} f_X(y), & |y| \leq A \\ 0 & \text{elsewhere} \end{cases}$$

$$f_Y(y) = p_1 \delta(y + A) + \tilde{f}_Y(y) + p_2 \delta(y - A)$$

Saturation Distorsion



$$\begin{aligned} P_S = E\{S^2\} &= \int_{-\infty}^{\infty} s^2 f_S(s) ds = \int_{-\infty}^0 s^2 f_X(A - s) ds + \int_0^{\infty} s^2 f_X(-A - s) ds \\ &= \int_A^{\infty} (x - A)^2 f_X(x) dx + \int_{-\infty}^{-A} (x + A)^2 f_X(x) dx \end{aligned}$$

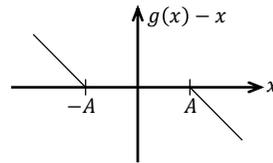
SDR – Signal-to-Distorsion Ratio

$$\text{SDR} = \frac{P_X}{P_S}$$

$$\text{SDR}_{\text{dB}} = 10 \cdot \log_{10} \left(\frac{P_X}{P_S} \right)$$

Saturation Error

$$S = Y - X = g(X) - X = \begin{cases} -A - X, & X < -A \\ 0, & |X| \leq A \\ A - X, & X > A \end{cases}$$

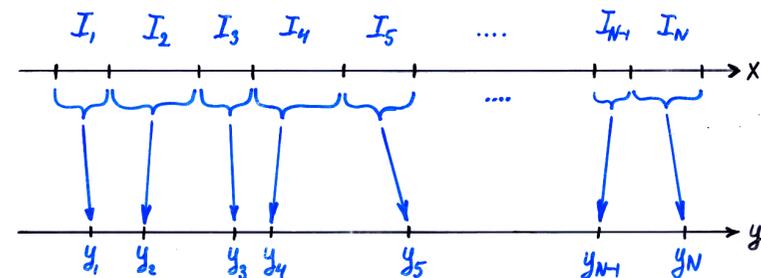
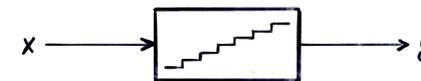


$$p_3 = \Pr\{|X| \leq A\} = \int_{-A}^A f_X(x) dx = 1 - p_1 - p_2$$

$$\tilde{f}_S(s) = \begin{cases} f_X(-A - s) & s > 0 \\ 0 & s = 0 \\ f_X(A - s) & s < 0 \end{cases}$$

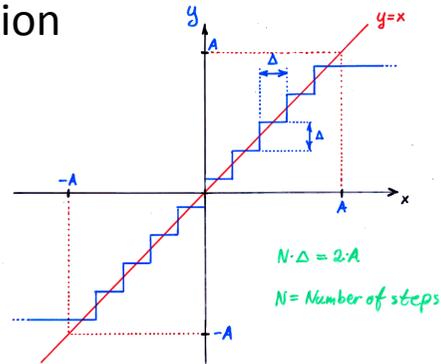
$$f_S(s) = \tilde{f}_S(s) + p_3 \delta(s)$$

Quantization Principles

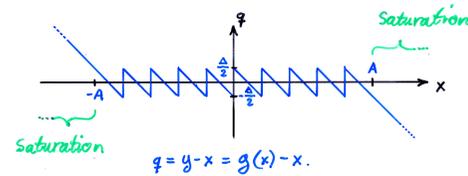


Uniform Quantization

$$y = g(x) = \begin{cases} A - \frac{\Delta}{2}, & x > A \\ \frac{\Delta}{2} + \lfloor \frac{x}{\Delta} \rfloor \cdot \Delta, & |x| \leq A \\ -A + \frac{\Delta}{2}, & x < -A \end{cases}$$



Quantization error:



Quantization Distorsion 2(2)

$$P_Q = \int_{-A}^A (g(x) - x)^2 f_X(x) dx = \sum_{k=1}^N \int_{y_k - \frac{\Delta}{2}}^{y_k + \frac{\Delta}{2}} (y_k - x)^2 f_X(x) dx = \int_{-A/2}^{A/2} u^2 f_X(y_k + u) du \approx \sum_{k=1}^N \int_{-A/2}^{A/2} u^2 f_X(y_k) du$$

Δ small, u small

$$= \sum_{k=1}^N f_X(y_k) \int_{-A/2}^{A/2} u^2 du = \frac{\Delta^2}{12} \sum_{k=1}^N \Delta \cdot f_X(y_k) \approx \frac{\Delta^2}{12} \sum_{k=1}^N \Pr\{X \in I_k\} = \frac{\Delta^2}{12}$$

Pr{X ∈ I_k} = 1

Error distribution: Approx. uniformly distr. over $[-\frac{\Delta}{2}, \frac{\Delta}{2}]$

Generally without saturation:

$$P_Q \leq \frac{\Delta^2}{4} \text{ since } |Q| \leq \frac{\Delta}{2}.$$

Quantization Distorsion 1(2)



The error: $Q = Y - X = g(X) - X$

Quantization distortion:

$$P_Q = E\{Q^2\} = E\{g(X) - X\}^2 = \int_{-\infty}^{\infty} (g(x) - x)^2 f_X(x) dx$$

Assumptions:

1. No saturation: $f_X(x) = 0$ for $|x| \geq A$
2. Nice distribution: $f_X(x)$ continuous for $|x| < A$
3. Small Δ : $f_X(x)$ approx. const. in intervals of length Δ .

SDR for Uniform Quantization

Still limited to $[-A, A]$ and nice enough distribution.

$$P_Q = E\{Q^2\} = \int_{-\Delta/2}^{\Delta/2} q^2 \frac{1}{\Delta} dq = \frac{\Delta^2}{12} = \frac{A^2}{3N^2} \Rightarrow \text{SDR} = \frac{P_X}{P_Q} = \frac{3P_X}{A^2} N^2 = \frac{3P_X}{A^2} 2^{2n}$$

$$\text{SDR}_{\text{dB}} = 10 \log_{10}(\text{SDR}) = 10 \log_{10}\left(\frac{3P_X}{A^2}\right) + n \cdot 20 \log_{10}(2) \approx 10 \log_{10}\left(\frac{3P_X}{A^2}\right) + 6n.$$

Example: Uniform distribution over $[-A, A]$.

$$P_X = E\{X^2\} = \int_{-A}^A x^2 \frac{1}{2A} dx = \frac{A^2}{3},$$

$$\text{SDR}_{\text{dB}} \approx 10 \log_{10}\left(\frac{3A^2/3}{A^2}\right) + 6n = 10 \log_{10}(1) + 6n = 6n$$

SDR for Uniform Quantization with Saturation

Uniform distribution over $[-B, B]$.

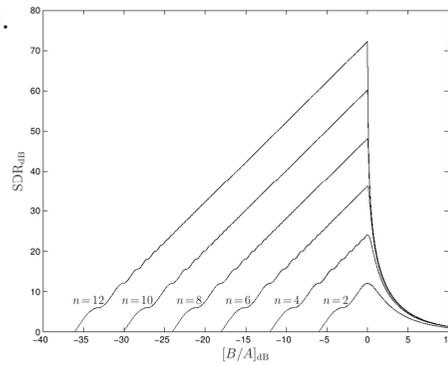
$$P_X = E\{X^2\} = \int_{-B}^B x^2 \frac{1}{2B} dx = \frac{B^2}{3}.$$

Q and S uncorrelated:

$$P_{Q+S} = P_Q + P_S.$$

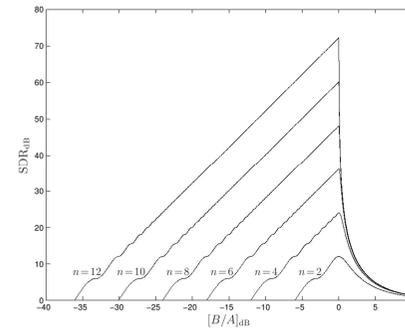
$$\text{SDR} = \frac{P_X}{P_Q + P_S},$$

$$P_{Q+S} = \begin{cases} \frac{(2k+1)(\Delta/2)^3 + (B-(2k+1)\Delta/2)^3}{3B}, & k\Delta \leq B < (k+1)\Delta, \quad k \in \{0, 1, \dots, N/2 - 2\} \\ \frac{A-\Delta/2}{B} \cdot \frac{\Delta^2}{12} + \frac{(B-A+\Delta/2)^3}{3B}, & B \geq A - \frac{\Delta}{2} \end{cases}$$

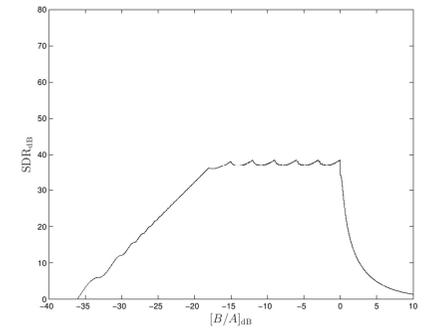


SDR for Non-Uniform Quantization

Uniform Quantization

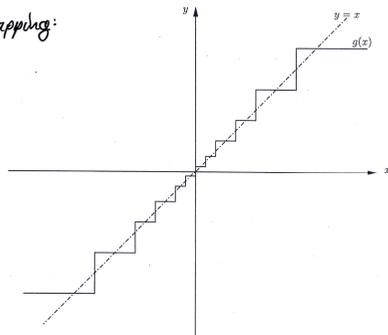


Non-Uniform Quantization

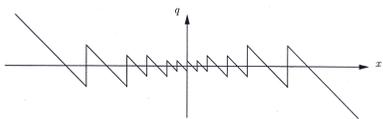


Non-Uniform Quantization

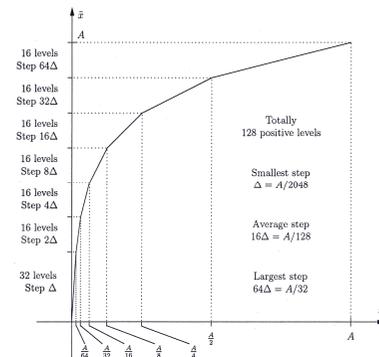
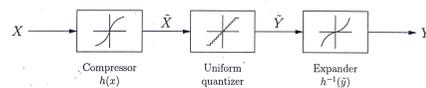
Mapping:



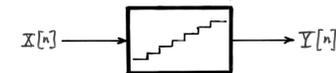
Quantization error:



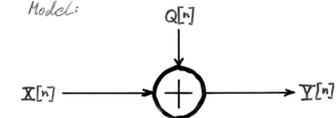
Realization



Modelling Quantization of a Stochastic Process



Model:



Model:

- The quantization noise is white
- The input and the quantization noise are uncorrelated.

Reason:

- It is almost true under reasonable assumptions.
- It gets more true with smaller quantization step.

The Quantization Noise is Almost White 1(4)

Quantization Noise: $Q[n] = Y[n] - X[n]$.

Model of PSD: $R_Q[\theta] = \frac{\Delta^2}{12}$

Assumptions:

1. that $f_{X[0],X[k]}(x_0, x_k) = 0$ holds for $|x_0| \geq A$ and for $|x_k| \geq A$,
2. that $f_{X[0],X[k]}(x_0, x_k)$ is three times differentiable inside the square $|x_0| < A, |x_k| < A$,
3. that $f_{X[0],X[k]}(-x_0, -x_k) = f_{X[0],X[k]}(x_0, x_k)$ holds.

Objective:

Show $\frac{r_Q[k]}{r_Q[0]} \rightarrow \delta[k]$, when $\Delta \rightarrow 0$,

The Quantization Noise is Almost White 3(4)

What about all those coefficients? Symmetry:

$$f_{X[0],X[k]}(-x_0, -x_k) = f_{X[0],X[k]}(x_0, x_k) \Rightarrow K_2 = K_3 = 0$$

Result:

$$r_Q[k] = \iint_{-\Delta/2}^{\Delta/2} q_0 q_k \frac{K_1 + K_4 q_0^2 + K_5 q_0 q_k + K_6 q_k^2 + K_7 (q_0^2 + q_k^2)^{3/2}}{\Delta^2} dq_0 dq_k.$$

Observation:

$$\iint_{-\Delta/2}^{\Delta/2} q_0 q_k dq_0 dq_k = \iint_{-\Delta/2}^{\Delta/2} q_0^3 q_k dq_0 dq_k = \iint_{-\Delta/2}^{\Delta/2} q_0 q_k^3 dq_0 dq_k = 0$$

Result:

$$r_Q[k] = \iint_{-\Delta/2}^{\Delta/2} \frac{K_5 q_0^2 q_k^2 + K_7 q_0 q_k (q_0^2 + q_k^2)^{3/2}}{\Delta^2} dq_0 dq_k.$$

The Quantization Noise is Almost White 2(4)

ACF of the Quantization Noise for $k \neq 0$:

$$r_Q[k] = \iint_{-\Delta/2}^{\Delta/2} q_0 q_k f_{Q[0],Q[k]}(q_0, q_k) dq_0 dq_k$$

2-D PDF of the Quantization Noise:

$$f_{Q[0],Q[k]}(q_0, q_k) = \begin{cases} \sum_{i_0=1}^{2A} \sum_{i_k=1}^{2A} f_{X[0],X[k]}(q_0 - A - \frac{\Delta}{2} + i_0 \Delta, q_k - A - \frac{\Delta}{2} + i_k \Delta), & |q_0| < \frac{\Delta}{2}, |q_k| < \frac{\Delta}{2}, \\ 0, & \text{elsewhere.} \end{cases}$$

2-D Taylor series expansion of $f_{X[0],X[k]}(q_0 - A - \frac{\Delta}{2} + i_0 \Delta, q_k - A - \frac{\Delta}{2} + i_k \Delta)$ gives us:

$$f_{Q[0],Q[k]}(q_0, q_k) = \frac{K_1 + K_2 q_0 + K_3 q_k + K_4 q_0^2 + K_5 q_0 q_k + K_6 q_k^2 + K_7 (q_0^2 + q_k^2)^{3/2}}{\Delta^2}$$

The Quantization Noise is Almost White 4(4)

We had:

$$r_Q[k] = \iint_{-\Delta/2}^{\Delta/2} \frac{K_5 q_0^2 q_k^2 + K_7 q_0 q_k (q_0^2 + q_k^2)^{3/2}}{\Delta^2} dq_0 dq_k.$$

Upper bound:

$$|r_Q[k]| < K \Delta^4 \Rightarrow \frac{|r_Q[k]|}{r_Q[0]} < 12K \Delta^2.$$

Result:

$$\frac{r_Q[k]}{r_Q[0]} \rightarrow \delta[k], \text{ when } \Delta \rightarrow 0.$$

Conclusion:

Almost white. Closer to white as Δ decreases.

Input & Quantization Noise Almost Uncorrelated

Same assumptions:

1. that $f_{X[0],X[k]}(x_0, x_k) = 0$ holds for $|x_0| \geq A$ and for $|x_k| \geq A$,
2. that $f_{X[0],X[k]}(x_0, x_k)$ is three times differentiable inside the square $|x_0| < A, |x_k| < A$,
3. that $f_{X[0],X[k]}(-x_0, -x_k) = f_{X[0],X[k]}(x_0, x_k)$ holds.

Normalized cross-covariance:

$$\rho_{XQ}[k] = \frac{\text{Cov}\{X[0], Q[k]\}}{\sigma_X \sigma_Q} = \frac{E\{(X[0] - m_X)(Q[k] - m_Q)\}}{\sigma_X \sigma_Q}$$

Objective:

Show $\rho_{XQ}[k] \rightarrow 0$, when $\Delta \rightarrow 0$, for all k .

Quantization – ACF & PSD Relations

Assumptions:

- $Q[n]$ is uniformly distributed on $[-\Delta/2, \Delta/2)$.
- $Q[n]$ is a white process.
- $Q[n]$ and $X[n]$ are uncorrelated.

ACF of output:

$$\begin{aligned} r_Y[k] &= E\{Y[n]Y[n+k]\} = E\{(X[n] + Q[n])(X[n+k] + Q[n+k])\} \\ &= E\{X[n]X[n+k]\} + E\{X[n]Q[n+k]\} + E\{Q[n]X[n+k]\} + E\{Q[n]Q[n+k]\}. \end{aligned}$$

Uncorrelated processes:

$$E\{X[n]Q[n+k]\} = E\{Q[n]X[n+k]\} = m_X m_Q = 0,$$

Result:

$$r_Y[k] = r_X[k] + r_Q[k]. \quad R_Y[\theta] = R_X[\theta] + R_Q[\theta].$$

Modelling Quantization Noise

Observation:

$$f_{X[0],X[k]}(-x_0, -x_k) = f_{X[0],X[k]}(x_0, x_k)$$

$$\Rightarrow f_X(x) \text{ \& } f_Q(q) \text{ even} \Rightarrow m_Q = m_X = 0$$

$$\Rightarrow \rho_{XQ}[k] = \frac{r_{XQ}[k]}{\sigma_X \sigma_Q} = \frac{E\{X[0]Q[k]\}}{\sigma_X \sigma_Q}$$

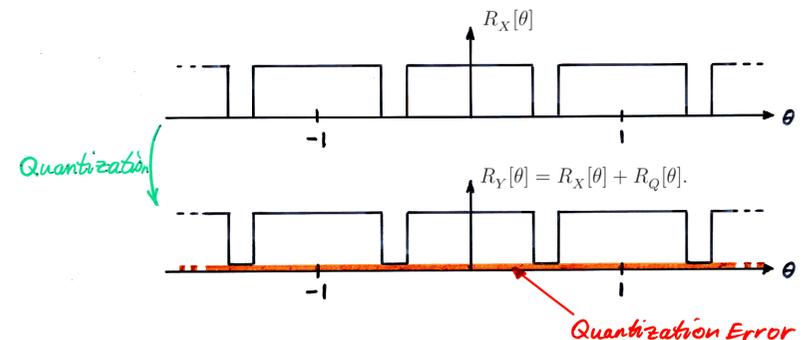
Similar reasoning as before:

$$|\rho_{XQ}[k]| < \sqrt{12}K\Delta/\sigma_X \Rightarrow \rho_{XQ}[k] \rightarrow 0, \text{ when } \Delta \rightarrow 0, \text{ for all } k.$$

Conclusion:

Almost uncorrelated. Less correlated as Δ decreases.

Quantization – Power-Spectral Densities



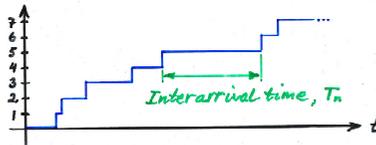
Poisson Processes 1(2)

A counting process, $X(t)$. Time continuous & amplitude discrete.
Counting the number of arrivals so far.

Examples:

- Customers entering a shop.
- Cars passing by
- Radio-active decay
- Light (counting photons)
- Packets in a network

A realization:



Properties of a stationary Poisson process:

$$X(0) = 0$$

$$X(t_1) \leq X(t_2) \text{ if } t_1 \leq t_2$$

$$\Pr\{X(t) = k\} = \frac{(\lambda t)^k}{k!} e^{-\lambda t}, \quad t \geq 0, k \in \mathbb{N}$$

$$\Pr\{X(t+\tau) - X(t) = k\} = \frac{(\lambda \tau)^k}{k!} e^{-\lambda \tau}, \quad t \geq 0 \& \tau \geq 0, k \in \mathbb{N}$$

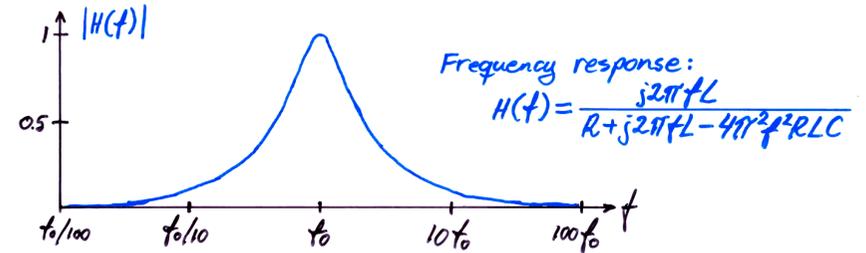
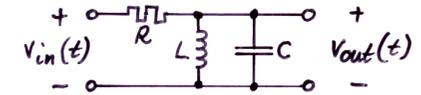
Interarrival times:

$$T_m \& T_n \text{ indep. for } m \neq n$$

$$f_{T_n}(t_n) = \lambda \cdot e^{-\lambda t_n}, \quad t_n \geq 0$$

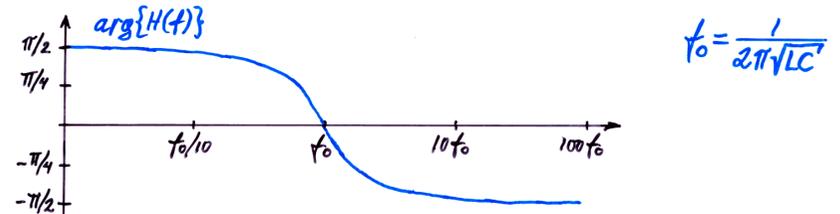
$$F_{T_n}(t_n) = 1 - e^{-\lambda t_n}, \quad t_n \geq 0$$

A Simple BP Filter



Frequency response:

$$H(f) = \frac{j2\pi fL}{R + j2\pi fL - 4\pi^2 f^2 RLC}$$



$$f_0 = \frac{1}{2\pi\sqrt{LC}}$$

Poisson Processes 2(2)

Expectation: $E\{X(t)\} = \sum_{k=0}^{\infty} k \Pr\{X(t) = k\} = \lambda t$ (λ intensity)

Variance: $\text{Var}\{X(t)\} = \lambda t$ for $t \geq 0$

Power: $E\{X^2(t)\} = E^2\{X(t)\} + \text{Var}\{X(t)\} = \lambda t(1 + \lambda t)$

Increments: $X(t+\tau) - X(t)$. Nonoverlapping increments are independent.

$$\text{ACF } (0 \leq t_1 \leq t_2): r_X(t_1, t_2) = E\{X(t_1)X(t_2)\} = E\{X(t_1)(X(t_2) - X(t_1) + X(t_1))\}$$

$$= E\{(X(t_1) - X(0))(X(t_2) - X(t_1))\} + E\{X^2(t_1)\}$$

Indep. incr. 0 ≤ t ≤ t₂

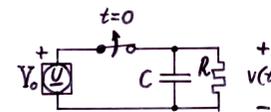
$$= E\{X(t_1) - X(0)\} \cdot E\{X(t_2) - X(t_1)\} + E\{X^2(t_1)\}$$

$$= \lambda t_1(\lambda t_2 - \lambda t_1) + \lambda t_1(1 + \lambda t_1) = \lambda t_1(1 + \lambda t_2)$$

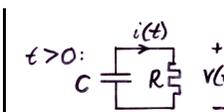
ACF $(0 \leq t_2 \leq t_1)$: $r_X(t_1, t_2) = \lambda t_2(1 + \lambda t_1)$ (similarly)

ACF (total): $r_X(t_1, t_2) = \lambda \cdot \min\{t_1, t_2\} + \lambda^2 t_1 \cdot t_2$

Discharging a Capacitor



Initial condition
 $v(0) = V_0$



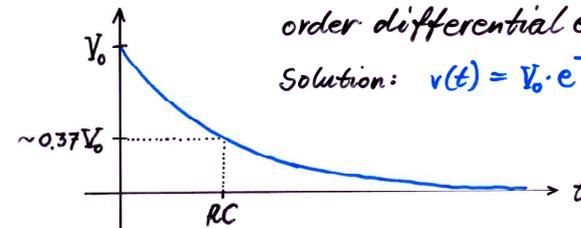
$$v(t) = R i(t) = -RC \frac{dv(t)}{dt}$$

$$i(t) = -C \frac{dv(t)}{dt}$$

$$\Rightarrow RC \frac{dv(t)}{dt} + v(t) = 0$$

This is the standard example of a first order differential equation.

Solution: $v(t) = V_0 \cdot e^{-t/RC}$ for $t \geq 0$



Amplitude Modulation

- The first technique used for radio broadcasts.
- A linear modulation technique.
- Simple to analyze.
- Simple demodulation.
- Noise sensitive.

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Amplitude Modulation – Deterministic Case

Standard AM:

$$x(t) = A \cdot (C + m(t)) \cos(2\pi f_c t)$$

Crystal receiver, an envelope detector,
first demodulator of standard AM:

